$\underline{\mathbf{Pr}~9.7}$  (a) If  $T_s=0.1$  the discrete-time signal is

$$x(0.1n) = [u(t) - u(t-1)] \mid_{t=0.1n} = \begin{cases} 1 & 0 \le 0.1n \le 1 \text{ or } 0 \le n \le 10 \\ 0 & \text{otherwise} \end{cases}$$

- (b) Expressing x[n] as indicated, then N = 11.
- (c) The Laplace transform of the sampled signal is

$$X_s(s) = \sum_{n=0}^{10} \mathcal{L}[\delta(t - nT_s)]$$
  
= 
$$\sum_{n=0}^{10} e^{-0.1ns}$$
  
= 
$$\frac{1 - e^{-1.1s}}{1 - e^{-0.1s}}$$

(d) The z-transform of the discrete-time signal is

$$X(z) = \sum_{n=0}^{10} z^{-n} = \frac{1 - z^{-11}}{1 - z^{-1}}$$

(e) To transform  $X_s(s)$  into X(z) we let  $z = e^{0.1s}$  .

 $\underline{\mathbf{Pr} \ 9.10}$  (a) The given signal can also be written

$$x[n] = \begin{cases} 1 & n \ge 0 \text{ and even} \\ 0 & \text{otherwise} \end{cases}$$

,

(b) Using the above expression for x[n], we have

$$X(z) = \sum_{n=0, \text{ even}}^{\infty} 1 z^{-n}$$
  
= 
$$\sum_{m=0}^{\infty} 1 z^{-2m} = \frac{1}{1 - z^{-2}} \qquad |z| > 1$$

where we let n = 2m to find the final expression.

(c) The z-transform of x[n] is also obtained by using its linearity

$$\begin{split} X(z) &= 0.5\mathcal{Z}[u[n]] + 0.5\mathcal{Z}[(-1)^n u[n]] \\ &= \frac{1}{2(1-z^{-1})} + 0.5\sum_{n=0}^{\infty} (-z^{-1})^n \\ &= \frac{1}{2(1-z^{-1})} + \frac{1}{2(1+z^{-1})} \\ &= \frac{1}{1-z^{-2}} \qquad |z^{-1}| < 1 \quad \text{or} \quad |z| > 1 \end{split}$$

(c) To find the poles and zeros let

$$X(z) = \frac{z^2}{z^2 - 1}$$

with poles  $z = \pm 1$ , and zeros z = 0, double.

Pr 9.11 (a) The Z-transform of the difference equation

$$y[n] = x[n] - 0.5y[n-1]$$
  $n \ge 0$ 

with initial condition y[-1] is

$$Y(z) = X(z) - 0.5(z^{-1}Y(z) + y[-1])$$

so that

$$Y(z) = \frac{X(z)}{1 + 0.5z^{-1}} - \frac{0.5y[-1]}{1 + 0.5z^{-1}}$$

(b) If X(z) = 1, y[-1] = 2 then Y(z) = 0 and therefore y[n] = 0 for  $n \ge 0$  but y[-1] = 2. If X(z) = 1 or  $x[n] = \delta[n]$  and y[-1] = 2 the difference equation is

$$y[n] = \delta[n] - 0.5y[n-1]$$
  $n \ge 0$ 

and can be solved recursively

$$y[0] = 1 - 0.5 \times 2 = 0$$
  

$$y[1] = 0 - 0.5 \times 0 = 0$$
  

$$y[2] = 0 - 0.5 \times 0 = 0$$
  

$$\vdots$$

(c) If y[-1] = 0 and  $x[n] = \delta[n]$  then we can compute y[n] = h[n], i.e., the impulse response. The corresponding transfer function is then from the equation for Y(z):

$$H(z) = \mathcal{Z}[h[n]] = \frac{Y(z)}{X(z)} = \frac{1}{1 + 0.5z^{-1}}$$

If we want  $y[n]=\delta[n]+0.5\delta[n-1]$  or  $Y(z)=1+0.5z^{-1}$  then

$$X(z) = \frac{Y(z)}{H(z)} = (1 + 0.5z^{-1})^2 = 1 + z^{-1} + 0.25z^{-2}$$

which gives  $x[n] = \delta[n] + \delta[n-1] + 0.25\delta(n-2)$ 

**<u>Pr 9.17</u>** (a) The signal  $x[n] = \delta[n] + \delta[n-1] + \delta[n-2]$  has a Z-transform

$$X(z) = 1 + z^{-1} + z^{-2}$$

(b) Then

$$Y(z) = X^{2}(z) = (1 + z^{-1} + z^{-2})^{2} = 1 + 2z^{-1} + 3z^{-2} + 2z^{-3} + z^{-4}$$

The convolution of the coefficients of X(z), or x[n], with themselves gives the sequence

$$y[n]=\delta[n]+2\delta[n-1]+3\delta[n-2]+2\delta[n-3]+\delta[n-4]$$

The length of y[n] is twice that of x[n] minus one, or  $2 \times 3 - 1 = 5$  so that Y(z) is a fourth -degree polynomial. The above result is verified using MATLAB

```
%% Pr 9.17
x=[1 1 1];
y=conv(x,x); N=length(y);
x1=[x zeros(1,N-3)]
n=0:N-1;
figure(1)
subplot(211)
stem(n,x1),ylabel('x[n]');grid
subplot(212)
stem(n,y);ylabel('y[n]'); grid
```

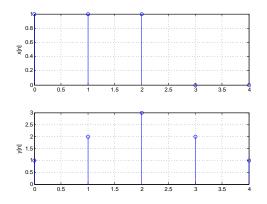


Figure 9.4: .

**<u>Pr 9.18</u>** Writing X(z) using terms found in tables, its partial fraction expansion is

$$X(z) = \frac{2 - z^{-1}}{2(1 + 0.25z^{-1})(1 + 0.5z^{-1})}$$
$$= \frac{A}{1 + 0.25z^{-1}} + \frac{B}{1 + 0.5z^{-1}}$$

corresponding to the poles at -0.25 and -0.5. The coefficients of the expansion are

$$A = \frac{2 - z^{-1}}{2(1 + 0.5z^{-1})} |_{z^{-1} = -4} = -3$$
$$B = \frac{2 - z^{-1}}{2(1 + 0.25z^{-1})} |_{z^{-1} = -2} = 4$$

so that

$$X(z) = \frac{-3}{1 + 0.25z^{-1}} + \frac{4}{1 + 0.5z^{-1}}$$

and the inverse is

$$x[n] = [-3(-0.25)^n + 4(-0.5)^n]u[n]$$

and in the steady-state it is zero.

**Pr 9.19** (a) F(z) is a proper rational function in positive powers of z as its numerator is of lower order than the denominator. If we convert it into negative powers,  $z^{-1}$ , we have

$$F(z) = \frac{z^{-2}(1+z^{-1})}{1-z^{-1}}$$

which is not proper rational in  $z^{-1}$  as its numerator is of higher order than its denominator. (b) Using the above expression we have that

$$F(z) = \frac{z^{-2}}{1 - z^{-1}} + \frac{z^{-3}}{1 - z^{-1}}$$

which gives

$$f[n] = u[n-2] + u[n-3]$$

given that  $1/(1-z^{-1})$  is the Z-transform of u[n] and  $z^{-2}$  and  $z^{-3}$  delay u[n] by 2 and 3 samples.