Signals and Systems Homework 4

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Problem 1

Determine the Laplace transform of each of the following signals and specify the corresponding regions of convergence.

- (1) $f(t) = e^{-a|t|}$
- (2) $f(t) = |t|e^{-a|t|}$
- (3) $f(t) = \sin(w|t|)$
- (4) $f(t) = \cos(w|t|)$
- (5) $f(t) = e^{-a|t|} \cos(w|t|)$
- (6) $f(t) = e^{-a|t|+jw|t|}$
- (7) $f(t) = |t| \sin(w|t|)$
- (8) $f(t) = |t|^n e^{-a|t|}$, n is positive integer.

Solution

(1) $\mathcal{L}(s) = \frac{1}{a-s} + \frac{1}{a+s}$ ROC: $-a < \mathcal{R}e[s] < a$ (2) $\mathcal{L}(s) = \frac{1}{(a+s)^2} + \frac{1}{(a-s)^2}$ ROC: $-a < \mathcal{R}e[s] < a$ (3) it doesn't converges (4) it doesn't converges (4) It doesn't converges (5) $\mathcal{L}(s) = \frac{s+a}{(a+s)^2+w^2} + \frac{a-s}{(a-s)^2+w^2}$ ROC: $-a < \mathcal{R}e[s] < a$ (6) $\mathcal{L}(s) = j[\frac{w}{(a+s)^2+w^2} + \frac{w}{(a-s)^2+w^2}] + \frac{s+a}{(a+s)^2+w^2} + \frac{a-s}{(a-s)^2+w^2}$ ROC: $-a < \mathcal{R}e[s] < a$ (7) it doesn't converges (8) $\mathcal{L}(s) = \frac{n!}{(a+s)^{n+1}} + \frac{n!}{(a-s)^{n+1}}$ ROC: $-a < \mathcal{R}e[s] < a$

Problem 2

Determine the Laplace transform of each of the following signals and specify the corresponding regions of convergence ,then plot the poles and zeros.

(1)
$$x(t) = e^{-4|t|} + e^{-5t} \sin(5t)u(t)$$

(2) $x(t) = \int_{-\infty}^{t} (e^{-2\tau} - 1)u(\tau)d\tau + e^{-3t}u(t)$
(3)

 $x(t) = \begin{cases} t & 0 \le t \le 1\\ 2-t & 1 \le t \le 3 \end{cases}$

Solution

(1) Using an approach similar to Problem 1, we have

$$e^{-4|t|} \xrightarrow{\mathcal{L}} \frac{1}{4-s} + \frac{1}{4+s} \operatorname{ROC} : -4 < \mathcal{R}e[s] < 4$$

and

$$e^{-5t}\sin(5t)u(t) \xrightarrow{\mathcal{L}} \frac{5}{(s+5)^2+25} \operatorname{ROC} :\mathcal{R}e[s] > -5$$

Thus

$$X(s) = \frac{1}{4-s} + \frac{1}{4+s} + \frac{5}{(s+5)^2 + 25} \ \operatorname{ROC} : -4 < \mathcal{R}e[s] < 4.$$



(2) We can obtain the following expressions easily:

$$\begin{split} (e^{-2t}-1)u(t) & \stackrel{\mathcal{L}}{\longrightarrow} \frac{1}{s+2} - \frac{1}{s} \ \operatorname{ROC} : \mathcal{R}e[s] > 0. \\ \int_{-\infty}^{t} (e^{-2\tau}-1)u(\tau)d\tau & \stackrel{\mathcal{L}}{\longrightarrow} \frac{\frac{1}{s+2} - \frac{1}{s}}{s} \ \operatorname{ROC} : \mathcal{R}e[s] > 0, \end{split}$$

and

$$e^{-3t}u(t) \xrightarrow{\mathcal{L}} \frac{1}{s+3} \operatorname{ROC} : \mathcal{R}e[s] > -3.$$

Therefore,

$$X(s) = \frac{-2}{(s+2)s^2} + \frac{1}{s+3} \text{ ROC } :\mathcal{R}e[s] > 0.$$



(3) It is not easy to obtain the Laplace transform of x(t) easily, so we use the property of Laplace transform: $\frac{d}{dt}x(t) = sX(s)$. Then the second derivative of x(t) is

$$\frac{d^2[x(t) - u(t-3)]}{dt^2} = \delta(t) + \delta(t-3) - 2\delta(t-1),$$

and the corresponding Laplace transform is

$$s^{2}[X(s) - \frac{e^{-3s}}{s}] = 1 + e^{-3s} - 2e^{-s}.$$

Therefore,

$$X(s) = \frac{1 + e^{-3s} - 2e^{-s}}{s^2} + \frac{e^{-3s}}{s}.$$

One zeros, $x \approx -0.422$.

Problem 3

A system is described by the following differential equation (see below). Find the expression for the transfer function of the system, Y(s)/X(s), assuming zero initial conditions.

$$\frac{d^3y}{dt^3} + 3\frac{d^2y}{dt^2} + \frac{dy}{dt} + y = \frac{d^3x}{dt^3} + 4\frac{d^2x}{dt^2} + 6\frac{dx}{dt} + 8x$$

Consider what if the initial conditions are not zero.

Solution

(a) If the initial condition is zero, then

$$\mathcal{L}(\frac{d^n y(t)}{dt^n}) = s^n Y(s)$$

We have

$$s^{3}Y(s) + 3s^{2}Y(s) + sY(s) + Y(s) = s^{3}X(s) + 4s^{2}X(s) + 6xX(s) + 8X(s)$$
$$Y(s)[s^{3} + 3s^{2} + s + 1] = X(s)(s^{3} + 4s^{2} + 6s + 8)$$
$$\frac{Y(s)}{X(s)} = \frac{s^{3} + 4s^{2} + 6s + 8}{s^{3} + 3s^{2} + s + 1}$$

(b) If the initial condition is non zeros, we assume the initial conditions are $y(0), y(0)^{(1)}, y(0)^{(2)}$, then

$$\mathcal{L}(\frac{d^n y(t)}{dt^n}) = s^n Y(s) - \sum_{r=0}^{n-1} s^{n-r-1} y(0)^r$$

We have

$$s^{3}Y(s) - (s^{2}y(0) + sy(0)^{(1)} + y(0)^{(2)}) + 3s^{2}Y(s) - 3(sy(0) + y(0)^{(1)}) + sY(s) - y(0) + Y(s)$$

= $s^{3}X(s) + 4s^{2}X(s) + 6xX(s) + 8X(s)$

Thus

$$Y(s) = \underbrace{\frac{s^3 + 4s^2 + 6s + 8}{s^3 + 3s^2 + s + 1}X(s)}_{ZSR} + \underbrace{\frac{(s^2 + 3s + 1)y(0) + (s + 3)y(0)^{(1)} + y(0)^{(2)}}{s^3 + 3s^2 + s + 1}}_{ZIR}$$

Problem 4

Let x(t) be the sampled signal specified as

$$x(t) = \sum_{n=0}^{\infty} e^{-nT} \delta(t - nT)$$

where T > 0

- (a) Determine X(s), including its region of convergence.
- (b) Sketch the pole-zero plot for X(s)
- (c) Use geometric interpretation of the pole-zero to argue that $X(j\Omega)$ is periodic.

Solution

(a) Note that





In order to determine the ROC, let us first find the poles of X(s). Clearly , the poles occur when This implies that the poles satisfy the following equation :

$$e^{-T(1+s_k)} = e^{jk2\pi}, k = 0, \pm 1, \pm 2, \cdots$$

Taking the logarithm of both side of the above equation and simplifying , we get $s_k = -1 + \frac{jk2\pi}{T}$, $k = 0, \pm 1, \pm 2, \cdots$.

Therefore, the poles all lie on a vertical line (parallel to the jw-axis) passing though s = -1. Since the signal is right-sided, the ROC is Re{s}>-1.

(b)The pole-zero poles is as the figure.

(c)The magnitude of the Fourier transform $X(j\Omega)$ is given by the product of the reciprocal of the lengths of the vectors from the poles to the point jw .The phase of $X(j\Omega)$ is given by the negative of the sum of the angles of these vectors. Clearly from the pole-zero plot above it is clear that both the magnitude and phase have to vary periodically with a period of $\frac{2\pi}{T}$.

(d)

$$\begin{split} Y(s) &= \sum_{n=0}^{\infty} \cos(nT) \exp(-snT) \\ &= \frac{e^{T(j-s)} - e^{T(-j-s)}}{1 - e^{T(j-s)} - e^{T(-j-s)} + e^{-2Ts}} \ \text{ROC} : \mathcal{R}e[s] > 0. \end{split}$$

No zeros, poles: $s = \pm j + j \frac{2k\pi}{T}$.